

Design and Implementation for optical fiber communication system using frequency shift coding

*Samar Y.Al-Dabagh**

*Hani J.Kbashi***

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Abstract:

In this research, optical communication coding systems are designed and constructed by utilizing Frequency Shift Code (FSC) technique. Calculations of the system quality represented by signal to noise ratio (S/N), Bit Error Rate (BER), and Power budget are done.

In FSC system, the data of Nonreturn-to-zero (NRZ) with bit rate at 190 kb/s was entered into FSC encoder circuit in transmitter unit. This data modulates the laser source HFCT-5205 with wavelength at 1310 nm by Intensity Modulation (IM) method, then this data is transferred through Single Mode (SM) optical fiber. The recovery of the NRZ is achieved using decoder circuit in receiver unit.

The calculations of BER and S/N for FSC system at maximum fiber length at 61.2 km equal to 2.30551×10^{-12} , 47.88526 dB respectively. The power budget for FSC system was calculated to be 29 dB.

Results show that the BER increases when the received optical power decreases due to increase of the optical fiber length 61.2 km. while S/N decreases. The optical power budget increases as the transmitted optical power increases.

Key words: optical fiber, coding, frequency shift, digital communication

Introduction:

Diverse communication and signal processing technologies utilize specially coded signal formats in order to achieve desirable capabilities such as error correction, interference rejection, and secrecy. In 1963, in an effort to standardize data communication codes, the United States adopted the bell system model 33 teletype code as the United States of America Standard Code for Information Interchange (USASII), better known simply as ASCII-63. ASCII is a 7-bit character set which has 128 combinations that means 7 bits in every block [1]

Since conventional NRZ code can prove unsuitable, especially for high data rate, for transmitting data in optical fiber, the best pattern for high

speed transmission is Manchester coding or frequency shift coding. Manchester transmission is alternate 1S and 0S while the best pattern for high speed frequency shift transmission is all Zeros, as shown in figure 1. FSC waveform shows that the levels change at the leading edge of each bit regardless of the bit value. For ones, encoder give an additional change at each bits center, zero have no second level change per bit[2].

The main purpose of Frequency Shift encoder is to convert the NRZ data waveform into FSC data waveform with the transition at center bit for each logic 1 bits from NRZ data, and in addition to transition at leading edge, for each 0 from NRZ data.

* College of Science for Women, Phys. Dept. Baghdad Un.

**College of Science, Phys. Dept., Baghdad Un.

There are three major multiple access approaches. Each user is allocated a specific frequency (wavelength) slot in Wavelength-Division Multiplexing (WDM). Both techniques have been extensively explored either using Wavelength-Division Multiplexing (WDM) or Time Division Multiplexing (TDM). Alternatively Optical Code Division Multiplexing (OCDM) is an alternative technique that takes the advantage of the enormous bandwidth in single mode fiber to achieve random asynchronous communication access among many users, free of network control [3]. Meagher B. al. design and demonstrated ultra low optical switching for WDM network [4]. Every user is assigned a specific code sequence. Through a proper choice of

the OCDM codes, the signals from all network nodes can be made mutually noninterfering in the incoherent OCDM. In encoder map each bit "1" of source information into a high bit rate optical sequence of ultra short light pulses, while the bit "0" is not encoded. D.E Leaird et al. have demonstrated spectrally phase coded OCDM with a modulation format based on switching between two codes [5]. A code switching security provides enhanced in OCDM and this technology could be utilized for a high capacity. Z. Jiang et al. have demonstrated a 2.5Gb/s four user, OCDMA system with a bit rate $< 10^{11}$ [6] and a wide band Frequency shift keying (FSK) have been demonstrated by Main and Chun z in 2006 [7]

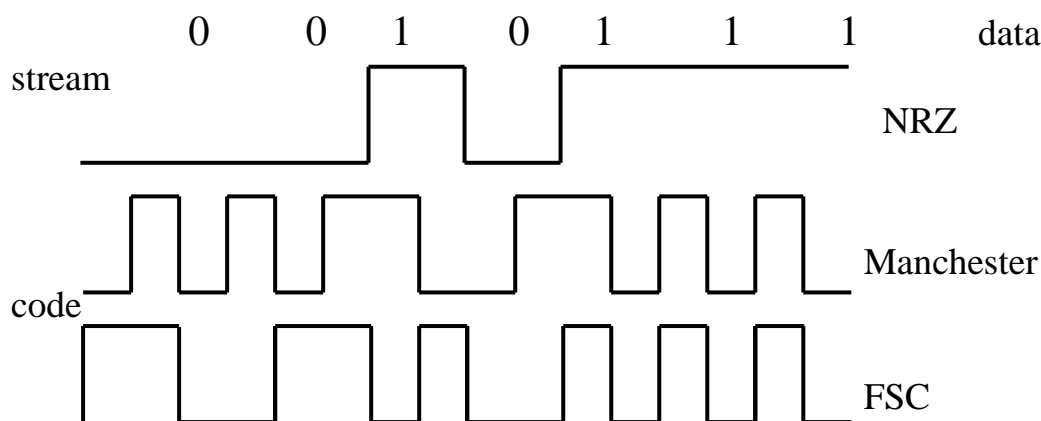


Fig.(1) Some methods of coding binary data

Materials and Methods:

Shift code system is shown in figure 2. This figure shows the principle components of the link which contains the transmitter (encoder and modulator), the receiver (demodulator and decoder), and the optical fiber as transmission channel for the link. A NRZ data rate at 190 kbits/s was entered into FSK encoder. This NRZ data that generated by voice communication represented by time division multiplexing system (TDM). The

optical source which provides the electrical-optical conversion is a semiconductor laser diode (LD). The package of this laser is designed to allow repeatable coupling into single mode optical fiber. Frequency shift code signal is used to drive an optical source (1310 nm) using Intensity Modulation (IM). This laser spectral width is 1nm, optical rise time is 2nsec, and the average optical output power is -10 dBm that was measured by using optical power meter OPM-4.

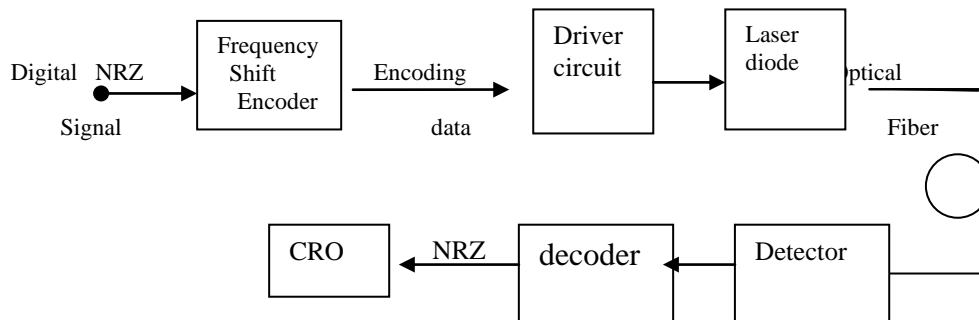


Fig.(2):General Block Diagram of Frequency Shift Coding System

The specification of the fiber that used in this system are core to cladding diameter equal to 10/100 micrometer , the whole all diameter is 3.2 mm, the attenuation is 0.2 dB for wavelength at 1310 nm , the numerical aperture is 0.14 and the acceptance angle is 8.05° .

The receiver consists of an optical detector which drives a future electrical stage and hence provides demodulation of the optical carrier. PIN photodetector was used in the receiver unit . Figure 3 shows the block diagram of the Frequency Shift decoder

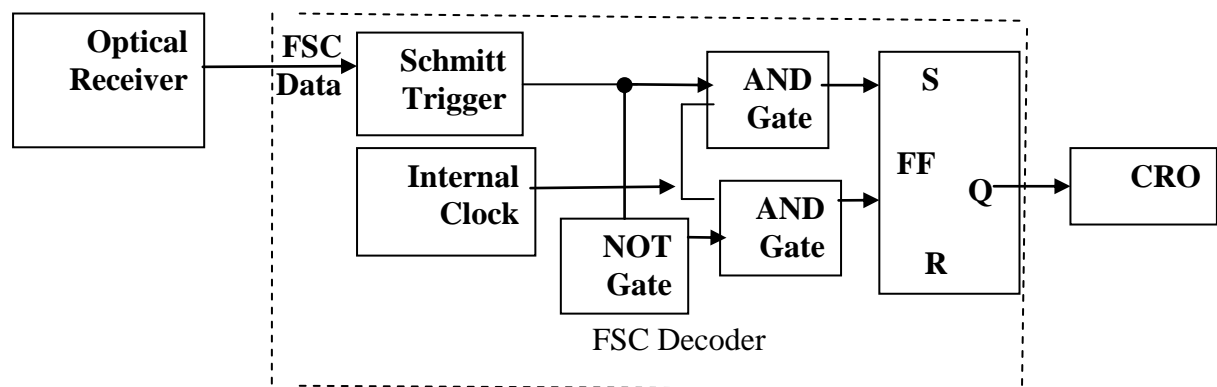


Fig.(3):Block diagram of the receiver unit of FSC .

Calculation and Results:

The waveforms from the Cathode Ray Oscilloscope (CRO)of the input NRZ data and the output data FSC of the transmitter unit are shown in figure 4.

The received pulses (FSC data) and the decoded NRZ signal waveforms are illustrated in figure5

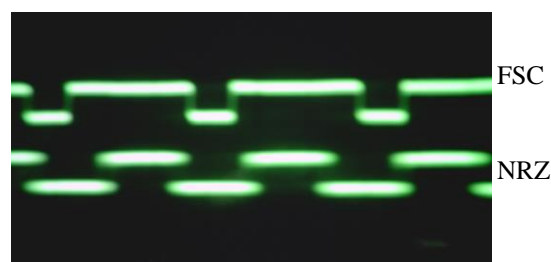


Fig.(4):the waveforms of NRZ and FSC signal with scale time set to 0.5 μ s/div .

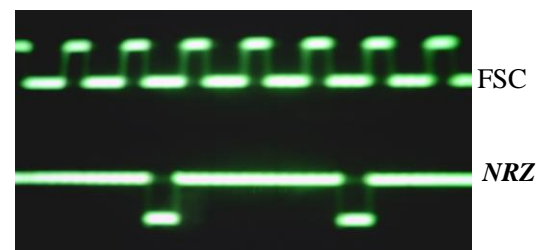


Fig. (5):The waveforms of FSC and NRZ signals of FSC decoder with scale time set to 0.5 μ s/div .

A-Computing of BER of FSC Receiver:

In order to determine the system quality, we calculate for the BER of the system in digital transmission. Is done. Performance is based on BER, which is the percentage of error bits received. BER as function of S/N of digital signal is given by[8]

$$BER = 1 / (6.28 S/N)^{1/2} \times EXP(-S/N / 2) \dots(1)$$

where S/N is defined as

$$S/N (dB) = 20 \log (I_s / I_n) \dots(2)$$

The signal current generated in the detector is I_s and is written as

$$I_s = P_{rec} \times R_\lambda \dots(3)$$

Where p_{rec} is the incident optical power .For HFCT - 5205 transceiver PIN photodiode ,the responsivity R_λ is equal to 0.36 A/W

and the total noise current I_n is written as

$$I_n = I_{th} + I_{shot} + I_d \dots (4)$$

The noise originating in the detector are thermal noise current (I_{th}), shot noise current (I_{shot}) and dark current(I_d) equal to 4 nA. The photodiode load resistor(R_L) 20 k Ω .

Calculations of BER based on a practical measurements of the received optical power at different values of the single mode optical fiber length of frequency shift code signal waveform are shown in table (1)

Table(1):The BER values as a function of fiber length, S/N, I_{th} , I_{shot} and p_{rec} for SM optical fiber channel with FSC system

$z(km)$	$p_{rec}(mw)$	$I_s(NA)$	$I_{th}(NA)$	$I_{shot}(NA)$	$I_n(NA)$	$S/N(dB)$	BER
1	0.1	36000	0.396636	1.47946	5.8761	75.74427	1.63563E-18
5.1	0.0733	26400	0.396636	1.26665	5.66328	73.36677	5.45606E-18
10.2	0.05978	21500	0.396636	1.14388	5.54052	71.78616	1.21572E-17
15.3	0.0348	12500	0.396636	0.872756	5.26939	67.52243	1.05679E-16
20.4	0.03195	11500	0.396636	0.836255	5.23289	66.84063	1.49362E-16
25.5	0.0223	8030	0.396636	0.698643	5.09528	63.94879	6.48338E-16
30.6	0.01989	7160	0.396636	0.659812	5.05645	63.02183	1.03814E-15
35.7	0.01592	5730	0.396636	0.590302	4.98694	61.20823	2.60866E-15
40.8	0.01216	4380	0.396636	0.515905	4.91254	58.9986	8.02078E-15
45.9	0.00837	3010	0.396636	0.428022	4.82466	55.91123	3.85746E-14
51	0.00706	2540	0.396636	0.393102	4.78974	54.49591	7.9287E-14
56.1	0.00409	1470	0.396636	0.299202	4.69584	49.92625	8.13781E-13
61.2	0.00321	1160	0.396636	0.265067	4.6617	47.88526	2.30551E-12

Table 1 shows the calculated results of the single mode optical fiber channel, which clarify noise current and data BER . The BER is increasing as the fiber length increasing . The acceptable order value of the BER in optical digital transmission is 10^{-9} .It is clear that the value of the BER of FSC system is equal to 2.355×10^{-12} at optical fiber length of 61.2 km is acceptable, then the received optical power at this length doesn't need any amplifying by optical amplifier

while S/N for FSC system at 61.2 km equal to 47.88526 dB .

From this table results, may conclude that the S/N is decreasing by increasing the fiber length as shown in figure 6. Although the S/N is decreasing ,the value of the S/N is acceptable for fiber length longer than 61.2 km.

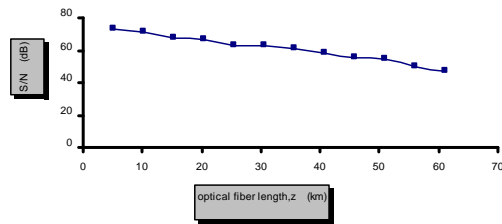


Fig.(6) :The relation ship between the Signal to Noise Ratio(S/N) with the optical fiber length ,Z in single mode optical fiber channel of Frequency shift code system.

The relationship between the BER with the received optical power as a function of fiber length of the SM optical fiber in FSC system is shown in figure 7

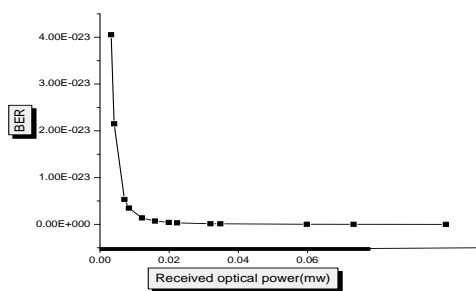


Fig. (7) :The relation ship between the logarithm of BER with the received optical power in single mode optical fiber channel of Frequency shift code system.

B-Optical Power Budget

To ensure that the optical fiber system has sufficient power for correct operation, the power budget must be calculated, the power budget is given by[9]

$$Power\ budget\ (dB) = 10\log\frac{P_{tx}}{P_{min}}$$

Or $Power\ budget\ (dB) = p_{tx}(dB) - p_{min}(dB)$
 ... (5)

$$Power\ budget\ (dB) = -10 - (-69) = 59\ dBm = 29\ dB$$

Where Ptx is the transmitted optical power from the laser diode. Pmin is the minimum detectable power at the receiver unit. From equation 5 the optical power budget equal 29dB and increases as the transmitted optical power increases.

Conclusion:

The encoded data stream shows successful demonstration of FSC generation techniques with high quality system of different fiber length(1-61.2) km., so FCC code technique is better suited for optical communication because of the large output signal to noise ratio S/N and low bit error rate BER of the received signal

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تصميم وبناء منظومة ترميز في أنظمة اتصالات الالياف البصرية باستخدام تقنية ازاحة التردد

هاني جاسم كباش**

سمر يونس الدباغ*

* جامعة بغداد كلية العلوم للبنات/قسم الفيزياء.
** جامعة بغداد/ كلية العلوم /قسم الفيزياء.

الخلاصة:

في هذا البحث، تم تصميم وبناء منظومة ترميز لمنظومة اتصالات الالياف البصرية باستعمال تقنية ترميز
ازاحة التردد (FSC) ثم حسب كل من معدل الخطأ (BER) في نقل البيانات ونسبة الاشارة الى الضوضاء
(S/N)، حساب ميزانية القدرة لمنظومة الاتصالات.
في تقنية ازاحة التردد (FSC) تم إدخال بيانات ترميز عدم العودة إلى الصفر (NRZ) وبمعدل
نقل 190kb/s إلى دائرة مرمز مانجستر (FSC Encoder)، عند وحدة الإرسال. عملت هذه الدائرة على
تحويل ترميز عدم العودة إلى الصفر إلى صيغة ترميز ازاحة التردد. هذه البيانات تعمل على تضمين
المصدر الليزري (520 - 5HFCT) ذو الطول الموجي 1310nm بطريقة تضمين الشدة (IM) ونقلت
بواسطة الليف البصري احادي النمط. تم استرداد البيانات المرمزة بترميز عدم العودة إلى الصفر باستعمال
دائرة فك الترميز (decoder circuit) عند وحدة الاستلام.
حسب كل من معدل الخطأ في نقل البيانات ونسبة الاشارة الى الضوضاء لمنظومة ترميز ازاحة التردد
اقصى طول لليف البصري المستخدم (61.2km) فكانت حوالي 2.30551×10^{-12} ، 47.88526 dB،
على التوالي تم حساب القدرة المتبقية كميزانية لمنظومة ازاحة التردد فكانت 29dB. اظهرت النتائج ان
معدل الخطأ في نقل البيانات يزداد عندما تقل القدرة المستلمة نتيجة زيادة طول الليف البصري (0-61.2)km
بينما نسبة الاشارة الى الضوضاء تقل عند نفس الطول. اما ميزانية القدرة لمنظومة الاتصالات البصرية
فتزداد مع زيادة القدرة المنتقلة.