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Mitigate PAPR Effect in Optical Orthogonal **Frequency Division Multiplexing Using Companding Technique**

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Abstract— orthogonal frequency division multiplexing (OFDM) is emerging as the leading technology for high data rates. The OFDM based on multicarrier transmission that transforms a wideband channel into a group of thin-band channels. The high peak to average power ratio (PAPR) considers the main drawback factor in this system, so the companding technique modifies the amplitude of OFDM signal to reduce this factor. The proposed A-law companding technique has been implemented in intensity modulation/direct detection optical orthogonal frequency division multiplexing (IM/DDcompanding OOFDM). The proposed scheme guarantees improved performance in terms of bit error rate (BER) and quality factor (OF) while reducing PAPR effectively and efficiently by modifying the amplitude of the transmitted signals. Our results confirm that the suggested scheme exhibits a good ability to reduce PAPR and a good BER performance based on the use the A factor. At the complementary cumulative distribution function (CCDF) of 10 ³, the PAPR value of our proposed scheme is about 4.4 dB lower than those of un-companded signal at best control factor used of A at 87.6, accordingly the OF is 2.5 dB and the BER is 6.2×10^{-3} at a distance 900km without any equalization technique or training sequence or any channel estimation.

Index Terms— Orthogonal Frequency Division Multiplexing (OFDM), Peak-to-Average Power Ratio (PAPR), Intensity Modulation/Direct Detection Optical Orthogonal Frequency Division Multiplexing (IM/DD OOFDM), complementary cumulative distribution function (CCDF).

I. INTRODUCTION

An OFDM has been used in the optical communication systems because of its opposition to Intersymbol Interference (ISI) and high spectral efficiency. IM/DD OOFDM system has been extensively explored in long-haul optical Communications systems [1-3].

As well as its many advantages, OOFDM systems have several disadvantages such as ISI and PAPR [4]. The high power amplifier working in the nonlinear section, therefore, In-band and out-of-band interferences are increased due to this nonlinearity. At last, the performance of the transmitter is enhanced by using PAPR reduction techniques [5]. Several methods have been investigated to decrease PAPR of OFDM signals such as clipping, pre-coding, companding, selective-mapping (SLM) and partial-transmission-sequence [6-7]. These techniques used to decrease PAPR, but there are some drawbacks. A clipping is the simplest one and is widely active for real-time implementation, but it causes additional clipping noise that degrades system's performance. The clipping causes both in-band distortion and out-of-band distortion [6], partial transmission technique [7-8], selective mapping [8-9] methods add extra complexity, coding complexity, or require extra optical components.

Due to increases in bandwidth and large memory with complexity in these techniques so the companding technique is useful for PAPR to overcome this constraint [9 -10].

The OOFDM signal involves severally modulated subcarriers that may provide a large PAPR once value-added up coherently. An OOFDM signal features a massive PAPR that is terribly sensitive to the non-linearity of the high peak [3]. Blocks of the symbol in the "OFDM" are selected with every symbol modulating one from a set of subcarriers and these subcarriers are recognized to be orthogonal [11].

The representation of complex OFDM signal is:

$$x(k) = \frac{1}{N} \sum_{n=0}^{N-1} X_n e^{\frac{j2\pi nk}{N}}$$
 (1)

Where
$$K=0, 1, 2, ..., N-1$$
 (2)

In this paper, due to large PAPR and low complexity, the propose A-law companding transform are using to mitigate the non-linearity that comes from PAPR by reduction this factor. The rest of the paper is organized as follows. In Section 2, a brief description of PAPR in an OFDM system. In Section 3, the definition of A-law companding reduction technique. In Section 4, descriptions of the proposed system, with A-law transform. In Section 5 results and discussion of the proposed system. in Section 6 conclusion with suggestions for further work.

II. PEAKS-TO-AVERAGE POWER RATIO (PAPR)

The PAPR for a signal x(K) is outlined as the ratio of maximum instantaneous power to the average power [10] as below:

$$PAPR = 10 * \log \left(\frac{\max|x(k)|^2}{Avg|x(k)|^2} \right)$$
(3)

Reduction of PAPR simply shows the probabilities that the PAPR of data block exceeds the given threshold value [11]. The expression of the CCDF is:

$$CCDF = Probability (PAPR > PAPR_{th})$$
(4)

Where PAPR_{th} represents the Threshold level.

III. A-LAW COMPANDING TECHNIQUE

Due to low complexes irrespective of the number of subcarriers, therefore the companding technique for PAPR reduction in OFDM systems is used [4]. A-law is done at the transmitter as a compressor after IFFT and an expander before FFT at the receiver [12].

The A-law has a non-zero value and it has mid risen at the origin point [12] as shown in Fig. 1. The high-level inputs the appearances are logarithmic segmented and for lowlevel inputs, the appearances are linearly segmented [13].

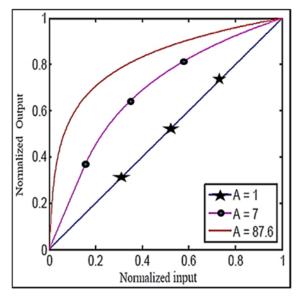


FIG. 1 CHARACTERISTICS OF A - LAW COMPRESSOR [5].

IV. THE PROPOSED MODELING SYSTEM OF COMPANDING TECHNIQUE IN IM/DD **OOFDM**

Firstly, the typical configuration of OOFDM system without using companded technique is showing in fig. 2 a generic OOFDM systems, containing 5 basic blocks: OFDM transmitter, optical upconverter, an optical link, optical downconverter, and OFDM receiver[14,15].

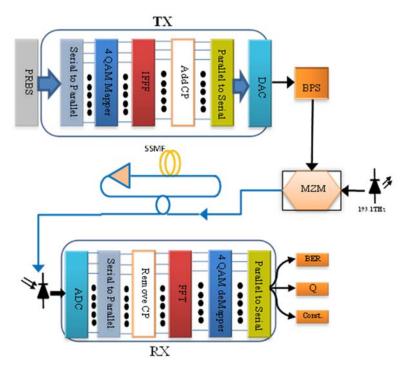


Fig. 2 The traditional OOFDM system.

For the proposed modeling system as compared with the traditional system, the A-Law transforms is applied after the IFFT at transmitter prior to before in receiver the as shown in fig. 3.

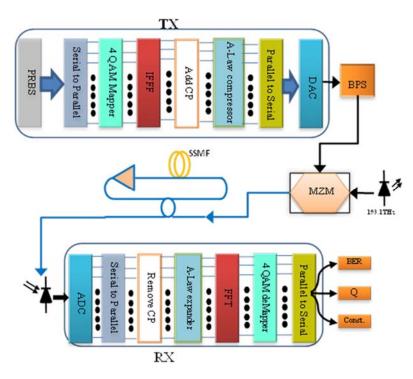


FIG. 3 OOFDM SYSTEM USING PROPOSED A- LAW COMPANDING TECHNIQUE.

On the transmitter side, the transmission bit rate is 10 Gbps. The bit streams are created using a pseudo-random binary sequence generator and the data are modulated by a 4QAM encoder. The information stream is further parsed into 128 parallel data subcarriers and processed by the IFFT processor. Accordingly, Table (1) summarizes the required parameter to implement the laser and channel performance.

TABLE .1. SIMULATION PARAMETERS FOR PROPOSED OOFDM

Global parameters		Optical Link	
BitRate	10 Gbps	SMF length	50 km
Sample Rate	40 Gbps	DCF length	10 km
Sequence length	8192	Number of loop	15
Bits per symbol	2	Attenuation for SMF	0.2 dB/km
OFDM modulator		Attenuation for DCF	0.5 dB/km
Max subcarriers	128	Dispersion SMF	16 ps /nm/km
Cyclic Prefix	1/8	Dispersion DCF	-80 ps /nm/km
CW Laser		CoreArea of SMF	$80~\mu m^2$
Carrier Frequency	193.1 THz	CoreArea of DCF	50 μm²
Power	10 mw	optical amplifier gain	12 dBm
Line Width	0.1 MHz	optical amplifier Noise figure	4 dBm

The OFDM data was produced and decoded by MATLAB, and VPI_Transmission_Maker did the up-down optical conversion and fiber link. At the receiver, the reverse operation is applicable to pick up the original signal that's transmitted before.

V. RESULTS AND DISCUSSION

The transmitted digital bit streams are originated in VPI_Transmission_Maker from the Pseudorandom binary sequence (PRBS) at (2¹³) and then OFDM data decoded with and without A-Law transform by Matlab software package, then optical modulation and optical up conversion, transmission and downconversion are done by VPI_Transmission_Maker.

The coded OFDM signal (real and imaginary part of companded OFDM) is up-converted at 7.5 GHz an intermediate frequency (IF) as shown in fig. 4.

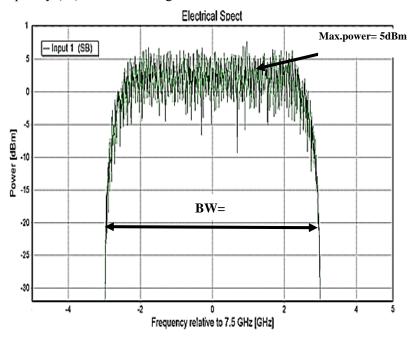


FIG. 4 PROPOSED ELECTRICAL OFDM SPECTRUM AT IF=7.5GHZ

The output spectrum power from MZM will contain an optical carrier which in this case is targeted at 193.1 THz with two side-bands targeted at 7.5 GHz of the optical OFDM signal as displayed in fig. 5.

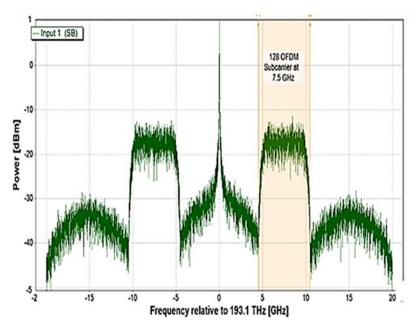


FIG. 5 THE SPECTRUM OF THE PROPOSED A-LAW OOFDM

Likewise, an optical filter is then responsible for suppressing the lower side-band by 18 GHz bandpass optical filter. The filtered OOFDM single shown in fig. 6.

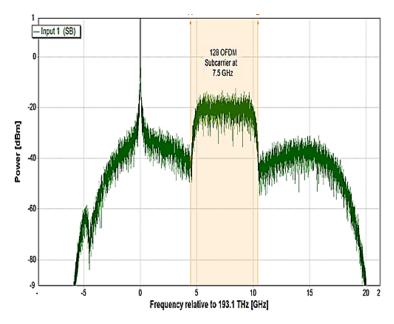


FIG. 6 THE SPECTRUM OF THE PROPOSED A-LAW OOFDM AFTER FILTERING.

Fig. 7 summarizes the PAPR performance of the proposed OOFDM transmitter is significantly more improved because of the controlling factor (A) where he is get increasing reach to an optimum value, the more compression added in the transmitter with reverses in receiver side to mitigate the effect of the PAPR. Therefore, the transmitted signals have uniform distribution with less power to save it from high dispersion that was resulted from traveling through link. The PAPR of the proposed technique at CCDF $\approx 10^{-3}$ is reduced by 2.2 dB at A=30 and 4.45 dB at A=87.6 as compared with the original signal.

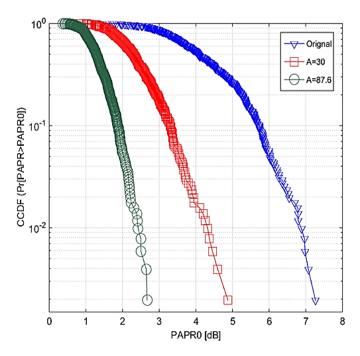


FIG. 7 CCDF PLOT IS SHOWING A-LAW PAPR REDUCTION

The propose A-Law transform algorithm has been clearly enhanced the performance of all transmitter systems of OOFDM as shown in the constellation diagram in fig. 8 with two value of factor A at 30 and 87.6.

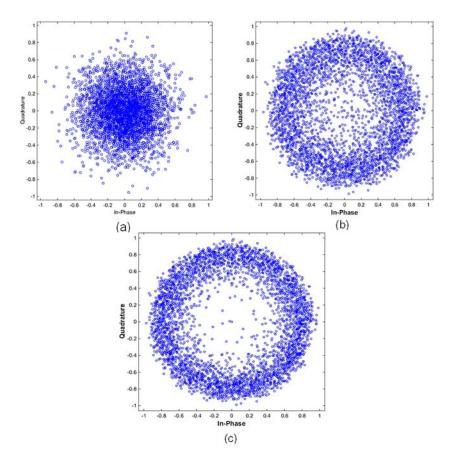


FIG. 8 CONSTELLATION DIAGRAM OF OOFDM AT END OF TX AT (A) ORIGINAL (B) PROPOSED COMPANDING AT A=30 (C) PROPOSED COMPANDING AT A=87.6

After the fiber length at 900 Km (60Km×15) with 4 QAM modulation format The QF, BER and constellation diagram at the end of the receiver are shown in fig. 9, fig. 10 and fig. 11 respectively. Moreover, the degradation observed in the original system without companded technique used is shown in fig. 9, fig. 10 plotted with black color and fig. 11(a), the measured QF is 2 dB, BER is 2.4×10^{-1} and non-uniform constellation diagram. The degradation is the maximum because of the increasing of the chromatic dispersion when increasing the transmission length.

Mostly, the proposed A-Law algorithm clearly improves the OOFDM system performance described as QF, BER, and constellation diagram that presented in fig.9, fig.10, and fig.11. In this system, proposed two values of A at 30 and 87.6, the A at 87.6 is totally enhancing the all system in term of QF at 2.5 dB and BER at 6.2×10^{-3} than that using A at 30 which QF = 1.4 dB and BER = 2.9×10^{-3}

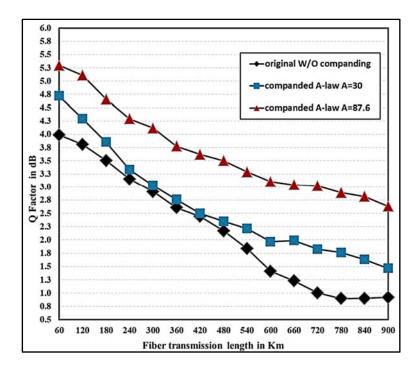


Fig. 9 QF vs Fiber Length

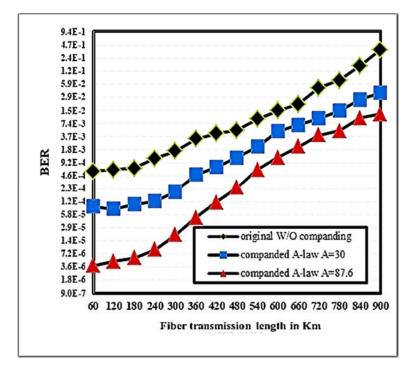


FIG. 10 BER VS FIBER LENGTH

Clearly as shown in fig. 11, the constellation performance is clearly the difference between the original and proposed technique. However, the control factor A at 87.6 a significant improvement in constellation performance over un-companded or companded at 30 of A.

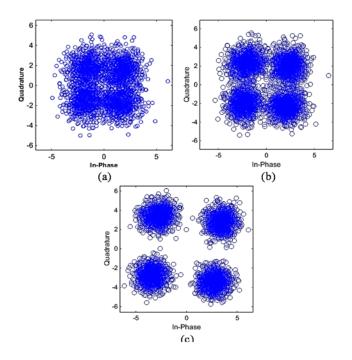


FIG. 11 CONSTELLATION DIAGRAM OF OOFDM SYSTEM AT THE RECEIVER END AFTER 900 KM AT (A) ORIGINAL (B) PROPOSED COMPANDING AT A=30 (C) PROPOSED COMPANDING AT A=87.6

VI. CONCLUSIONS

The proposed system in this paper has confirmed a defeat of fiber nonlinearity using A-law companding in intensity modulation/direct detection optical orthogonal frequency division multiplexing system to decrease the PAPR value of OOFDM systems. The proposed technique implemented without any equalization technique or training sequence or any channel estimation. The results show that the parameter A at 87.6 is a best one to reduce the PAPR and performance improvement of BER, QF and spectral efficiency more than that of A at 30. All simulation results have been presented by combination VPI Transmission Maker software with the Matlab software package. In future, the system performance can be more improvement by proposing another PAPR technique with the A-law companding to make hybrid with adjustable value of A.

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