

# EVOLUTION OF RANDOM VECTORS ALONG SINGLE-MODE FIBER IN PRESENCE OF CONVENTIONAL POLARIZATION EFFECTS

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## ABSTRACT

The presence of polarization mode dispersion (PMD) vector leads to differential group delay (DGD) among the polarization components, while the presence of polarization dependent loss (PDL) vector leads to attenuating one of the components and increases the other by a magnitude determined by PDL value. The study of each phenomenon individually does not give a proper description of the physical nature of the optical fiber system, because these two phenomena arise together at the same time. In this paper, we examine the combined PMD and PDL to generate random vectors at the concatenation sections as well as the total PMD vector. The derivation of the complex PMD vector formula of single section and concatenation sections leads to explain the properties of DGD and orthogonality of principal states of polarization (PSPs). However, new recursive formulas were found to the complex PMD vector, which are applied to illustrate the related random quantities.

Keywords: PMD, PDL and Polarization effects.

## الخلاصة

إن وجود ظاهرة تشتت نمط الاستقطاب (PMD) تؤدي إلى تأخير الزمرة التفاضلي (DGD) في مركبتي الاستقطاب، بينما وجود ظاهرة خسارة معتمدة الاستقطاب (PDL) تؤدي إلى توهين إحدى مركبتي الاستقطاب وزيادة المركبة الأخرى وحسب قيمة هذه الظاهرة. إن دراسة كل ظاهرة بشكل منفصل عن الأخرى لا يعطي وصفاً مناسباً للطبيعة الفيزيائية لنظام الليف البصري، لأن هاتين الظاهرتين تصلان معاً وفي نفس الوقت. في هذا البحث تم اختبار دمج الظاهرتين (PMD) و (PDL) لتوليد متجهات عشوائية خلال المقاطع المتسلسلة. إن اشتقاق صيغة متجه (PMD) المعقد لمقطع واحد ولعدد من المقاطع المتسلسلة يؤدي إلى فهم خصائص تأخير الزمرة التفاضلي (DGD) كذلك فهم تعامدية حالات الاستقطاب الأساسية (PSPs). تم إيجاد صيغة جديدة لمتجه (PMD) المعقد والذي طبق لتوضيح الصفات العشوائية المتعلقة بهذا البحث.

## 1. Introduction

It is well known that at high data rates (typically  $>10$  Gbit/s) polarization effects can severely impair optical communication system performance. Conventional polarization effects include polarization mode dispersion (PMD), which causes the differential group delay (DGD) between the principle states of polarization (PSPs), and polarization dependent loss (PDL), which causes polarization dependent attenuation of the propagating signal [N. Gisin and B. Huttner, 1997; N. Gisin et al, 2000]. DGD is a time delay at a discrete frequency between the fastest and slowest modes of an optical signal. This randomly varying delay causes optical pulses to broaden and hence bit errors. PMD is the mean of DGD over all frequencies [B. Huttner et al, 2000]. State of polarization (SOP) change is caused by change of PMD and PDL [I. Yoon and B. Lee, 2004]. PMD is caused by birefringence on a fiber's core/cladding breaking the cylindrical symmetry [J. Gordon and H. Kogelnik, 2000]. The PMD describes the polarization dependence of the time delay of an optical pulse as it propagates along the fiber. High amounts of DGD can cause pulses to overlap in an optical communication system [C. D. Poole and R. E. Wagner, 1986].

The PDL on a linear scale is defined as the ratio between the maximum and minimum attenuation coefficient over all polarization states. While this quantity can be measured by using scrambling and recording the maximum and minimum attenuation it is usually measured using the Jones matrix or Muller matrix method [L. Chen L et al, 2007; M. Shtaif and O. Rosenberg, 2005]. PDL describes the polarization dependence of the optical attenuation for different states of polarization and can occur concurrently with PMD in fibers [Y. Li and A. Yariv, 2005; G.P. Agrawal, 2005]. PDL is a varying insertion loss arising from the dependence of a component's transmission coefficient on the SOP [M. Shtaif and O. Rosenberg, 2005].

In a complex system with; optical fibers, couplers, filters, multiplexers/ demultiplexers, variable optical attenuators, erbium doped optical amplifiers, and add/drop multiplexing switchers, the combination of PMD and PDL will lead to a complex PSP vector [G.P. Agrawal, 2005; D. S. Waddy et al, 2003]. The combined PMD-PDL interaction can further degrade the system performance. Due to the interference between the fast and slow modes, the interaction between these effects will make an optical system more complicated than PMD or PDL alone. It may result in anomalous dispersion and causes additional signal distortion [M. Wang et al, 2003]. It is known that this interaction causes the fast and slow PSP direction to become non-orthogonal [C. Xie and L. F. Mollenauer, 2003]. This non-orthogonality is related to the imaginary part of the PSP vector or Differential Slope Attenuation (DSA) [S. Yang et al, 2005].

In this paper, a new recursive formula is presented to determine the total PMD vector as a function of the local random vectors to the concatenation sections. Also, we are illustrated many treatments that may be used to study the related behavior of the random quantities in single mode fiber.

## 2. Basic Treatments

Linear transmission of light from the input to the output of an optical link can always be expressed in the form

$$|s(w)\rangle = T(w)|t\rangle \quad (1)$$

where  $w$  is the light frequency,  $|t\rangle$  and  $|s(w)\rangle$  are the input and output polarization Jones vectors and  $T(w)$  is the transmission matrix. Unlike the familiar case accounting only PMD, where the transmission matrix is unitary, here  $T(w)$  is a general matrix with no significant limitations. In principal, an optical system with polarization effects can be modeled as a concatenation of  $N$  segments. Each segment  $j$  have the PDL and PMD effects that describe by  $T_{PDLj}$  and  $T_{PMDj}$ , respectively, which are defined as [B. Huttner et al, 2000]

$$T_{PDLj} = e^{\frac{\vec{\alpha}_j \cdot \vec{\sigma}}{2}} = \cosh(\alpha_j / 2) + (\hat{\alpha}_j \cdot \vec{\sigma}) \sinh(\alpha_j / 2) \quad (2)$$

$$T_{PMDj} = e^{-i \frac{\omega \vec{\tau}_j \cdot \vec{\sigma}}{2}} = \cos(\omega \tau_j / 2) - i(\hat{\tau}_j \cdot \vec{\sigma}) \sin(\omega \tau_j / 2) \quad (3)$$

where  $T_{PDLj}$  is a Hermitian matrix, i.e.  $T_{PDLj} = T_{PDLj}^\dagger$ , which have the real eigenvalues  $e^{\pm \alpha_j / 2}$ . Note that, in this representation PDL matrix, the polarization component of field that is parallel  $\vec{\alpha}_j$  experiences a gain  $e^{\alpha_j}$ , but the anti-parallel component is

attenuated by  $e^{-\alpha_j}$ . The vector  $\vec{\alpha}_j = \alpha_j \hat{\alpha}_j$ , which is pointing in the direction of the least attenuated state of polarization and is called PDL vector, stands for the  $j$ th PDL segment with value expressed in dB by [G.P. Agrawal, 2005]

$$PDL(dB) = 10 \log_{10}(T_{\max}/T_{\min}) = 20 |\alpha_j| \log_{10}(e) \quad (4)$$

where  $T_{\max} = 1$  and  $T_{\min} = e^{-2\alpha}$  are the maximum and minimum transmission intensities through the system, The matrix  $T_{PMDj}$  is unitary, i.e.  $T_{PMDj}^\dagger = T_{PMDj}^{-1}$ . The vector  $\vec{\tau}_j$  represents the local PMD vector and  $\vec{\sigma}$  is the vector of Pauli spin matrices whose three components are [9]

$$\sigma_1 = \begin{bmatrix} 1 & 0 \\ 0 & -1 \end{bmatrix}, \quad \sigma_2 = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}, \quad \sigma_3 = \begin{bmatrix} 0 & -i \\ i & 0 \end{bmatrix} \quad (5)$$

Note that, for any three-dimensional vector  $\vec{m}$ , the product  $\vec{m} \cdot \vec{\sigma}$  is defined as  $m_1\sigma_1 + m_2\sigma_2 + m_3\sigma_3$ , which is a complex  $2 \times 2$  matrix.

The frequency domain evolution of the state vector may be determined by differentiating of Eq.(1) and using the fact  $|t\rangle = T^{-1}|s\rangle$  to yield

$$\frac{\partial |s\rangle}{\partial \omega} = T_w T^{-1} |s\rangle \quad (6)$$

Note that, since  $TT^{-1} = I$  then the differentiation leads to  $T_w T^{-1} + TT_w^{-1} = 0$ . But,  $(T_w T^{-1})^{-1} = TT_w^{-1}$ , such that  $T_w T^{-1} = -(T_w T^{-1})^{-1}$ . In general, any  $2 \times 2$  matrix  $M$  may be expanded in the form  $M = a_o I + \vec{a} \cdot \vec{\sigma}$  with  $a_o = \text{trace}(M)/2$  and  $\vec{a} = \text{trace}(\vec{\sigma}M)/2$ . Using this property, we may write  $T_w T^{-1} = a_o I + \vec{a} \cdot \vec{\sigma}$  and  $(T_w T^{-1})^{-1} = -a_o I - \vec{a} \cdot \vec{\sigma}$ . However, the equalization of  $T_w T^{-1}$  and  $-(T_w T^{-1})^{-1}$  will make  $a_o = 0$ . That is;  $T_w T^{-1}$  is a traceless matrix, which is a similar to the case of pure PMD. Accordingly, one may be demonstrated

$$T_w T^{-1} = -\frac{i}{2} \vec{W} \cdot \vec{\sigma} \quad (7)$$

for some complex vector  $\vec{W}$ . The vector  $\vec{W} = \vec{\Omega} + i\vec{\Lambda}$  is always called the complex PMD vector. Simply,  $\vec{\Omega}$  and  $\vec{\Lambda}$  vectors represent the PMD and PDL, respectively. The DGD is  $\tau = \text{Re} \sqrt{\vec{W} \cdot \vec{W}}$ , while Differential Slope Attenuation DSA is defined as  $\eta = \text{Im} \sqrt{\vec{W} \cdot \vec{W}}$ . It can be shown that the eigenvectors of the matrix  $2iT_w T^{-1} = \vec{W} \cdot \vec{\sigma}$  are the output principal states  $|p_{\pm}\rangle$  of the system (with the PSP definition according to the first order frequency-independent), no matter whether the PDL is zero or not. The eigenvalues of  $\vec{W} \cdot \vec{\sigma}$  are  $\chi$ , which in turns out to be a complex number in systems with PDL. Formally,  $\vec{W} \cdot \vec{\sigma} |p_{\pm}\rangle = \pm \chi |p_{\pm}\rangle$ . The eigenvectors are orthogonal in systems without PDL because the matrix  $\vec{W} \cdot \vec{\sigma}$  is Hermitian, whereas the Hermitian property is lost and in turn the orthogonality is not hold anymore in presence of PDL. This means that the PSPs in Stokes space  $\hat{p}_{\pm} = \langle p_{\pm} | \vec{\sigma} | p_{\pm} \rangle$  are no longer antipodal. Moreover, owing to the traceless of  $T_w T^{-1}$ , its two eigenvalues can be written as  $\chi = \pm(\tau + i\eta)$ .

Using the definitions  $\vec{W} = \vec{\Omega} + i\vec{\Lambda}$  and  $\chi = \pm\sqrt{\vec{W} \cdot \vec{W}}$ , one may be found

$$\chi = \pm\sqrt{\vec{W} \cdot \vec{W}} = \pm(m \cos \rho + im \sin \rho) \quad (8)$$

where

$$m = \sqrt{(\Omega^2 - \Lambda^2)^2 + (2\Omega \cdot \Lambda)^2} \quad (9a)$$

$$\rho = \frac{1}{2} \tan^{-1} \left( \frac{2\Omega \cdot \Lambda}{\Omega^2 - \Lambda^2} \right) \quad (9b)$$

$$\tau = \pm m \cos \rho \quad (9c)$$

$$\eta = \pm m \sin \rho \quad (9d)$$

$$|\chi|^2 = \sqrt{(\Omega^2 - \Lambda^2)^2 + (2\Omega \cdot \Lambda)^2} \quad (9e)$$

$$|\vec{W}|^2 = \Omega^2 + \Lambda^2 \quad (9f)$$

Using the definitions  $\hat{p}_{\pm} = \langle p_{\pm} | \vec{\sigma} | p_{\pm} \rangle$  and  $\vec{W} \cdot \vec{\sigma} | p_{\pm} \rangle = \chi | p_{\pm} \rangle$ , one may be found [M. A. Reimer, 2007]

$$\hat{p}_{\pm} = \pm \frac{2}{|\chi|^2 + |\vec{W}|^2} [\tau \vec{\Omega} + \eta \vec{\Lambda} \pm \vec{\Omega} \times \vec{\Lambda}] \quad (10)$$

The angle between the PSP's vectors in Stokes space, which represents the orthogonality, may be found using Eq.(10) as

$$\cos \psi = \hat{p}_{+} \cdot \hat{p}_{-} = \frac{4[\Omega^2 \Lambda^2 - (\Omega \Lambda \cos \phi)^2 - \tau^2 \Omega^2 - \eta^2 \Lambda^2 - 2\tau \eta \Omega \Lambda \cos \phi]}{[\Omega^2 + \Lambda^2 + \sqrt{(\Omega^2 - \Lambda^2)^2 + (2\Omega \Lambda \cos \phi)^2}]^2} \quad (11)$$

where  $\phi$  represents the angle between  $\vec{\Omega}$  and  $\vec{\Lambda}$ . The largest pulse spreading is defined as [B. Huttner et al, 2000]

$$\tau_{\max} = \frac{1}{2} \sqrt{\frac{\tau^2 + \gamma \eta^2}{1 - \gamma}} \quad (12)$$

where the overlapping factor  $\gamma$  is defined as

$$\gamma = \frac{|\vec{W}|^2 - |\chi|^2}{|\vec{W}|^2 + |\chi|^2} \quad (13)$$

Depending on Eqs.(11)-(13), there are many important results that may be specified as follows:

If  $\vec{\Omega}$  and  $\vec{\Lambda}$  are parallel or anti-parallel, then  $\vec{\Omega} \times \vec{\Lambda} = 0$  and  $|\vec{W}|^2 = |\chi|^2$ . In turn,  $\cos \psi = -(\tau \Omega \pm \eta \Lambda)^2 / (|\vec{W}|^2)^2$ ,  $\gamma = 0$ , and  $\tau_{\max} = \tau$ . That is; the parallel and anti-parallel cases will present a different orientation of the PSPs without holding the orthogonality.

If  $\vec{\Omega}$  and  $\vec{\Lambda}$  are orthogonal, then  $\vec{\Omega} \cdot \vec{\Lambda} = 0$ ,  $\eta = 0$  and  $\tau = \sqrt{\Omega^2 - \Lambda^2}$ . Consequently,  $\cos \psi = -(\Omega^2 - 2\Lambda^2) / \Omega^2$ . That is; the DAS will be zero and the DGD will be

maximum. Moreover, if we take  $|\vec{\Lambda}|=0$ , then the simplest case that makes

$(\cos\psi = -1)$ , i.e.  $\hat{p}_{\pm} = \pm\vec{\Omega}/\Omega$  which are orthogonal, will be deduced.

The absence of PMD effects will make  $\vec{\Omega} = 0$ ,  $\tau = 0$ , and  $\eta = 0$  and consequently  $\hat{p}_{\pm} = 0$ , which is a trivial case that means no PSPs states.

### 3. The Complex PMD Vector

According to Gisin and Huttner (see Eq.(11) in Ref. [N. Gisin and B. Huttner,1997]),

$T = T_{PMD}T_{PDL}$  that will make  $T_w T^{-1} = -i\vec{\tau} \cdot \vec{\sigma}/2$  for  $N=1$ . This means that the PDL effects are not contained in the result. In this model, we are adopted the form

$T = T_{PDL}T_{PMD}$  that makes  $T_w T^{-1} = -iT_{PDL}\vec{\tau} \cdot \vec{\sigma} T_{PDL}^{-1}/2$  for  $N=1$ .

#### 3.1 Isolated Section Model

For single section, depending on the result  $T_w T^{-1} = -iT_{PDL}\vec{\tau} \cdot \vec{\sigma} T_{PDL}^{-1}/2$ , using Eq.(7)

and the fact  $T_{PDL}^{-1} = e^{-\vec{\alpha} \cdot \vec{\sigma}/2}$ , yields

$$\begin{aligned}\vec{W} \cdot \vec{\sigma} &= T_{PDL}(\vec{\tau} \cdot \vec{\sigma})T_{PDL}^{-1} \\ &= [\cosh(\alpha/2) + (\hat{\alpha} \cdot \vec{\sigma})\sinh(\alpha/2)](\vec{\tau} \cdot \vec{\sigma})[\cosh(\alpha/2) - (\hat{\alpha} \cdot \vec{\sigma})\sinh(\alpha/2)]\end{aligned}\quad (11)$$

Using the identities [5]

$$(\vec{\alpha} \cdot \vec{\sigma})(\vec{\beta} \cdot \vec{\sigma}) = \vec{\alpha} \cdot \vec{\beta} + i\vec{\alpha} \times \vec{\beta} \cdot \vec{\sigma} \quad (12a)$$

$$(\vec{\beta} \cdot \vec{\sigma})(\vec{\alpha} \cdot \vec{\sigma}) = \vec{\alpha} \cdot \vec{\beta} - i\vec{\alpha} \times \vec{\beta} \cdot \vec{\sigma} \quad (12b)$$

$$(\vec{\alpha} \cdot \vec{\sigma})(\vec{\beta} \cdot \vec{\sigma})(\vec{\alpha} \cdot \vec{\sigma}) = [2(\vec{\alpha} \cdot \vec{\beta})\vec{\beta} - \alpha^2\vec{\beta}] \cdot \vec{\sigma} \quad (12c)$$

rearrangement the result, and separating the real and imaginary parts, one may be obtained

$$\vec{\Omega} = (\cosh^2(\alpha/2) + \sinh^2(\alpha/2))\vec{\tau} - 2\sinh^2(\alpha/2)(\hat{\alpha} \cdot \vec{\tau})\hat{\alpha} \quad (13)$$

$$\vec{\Lambda} = 2\cosh(\alpha/2)\sinh(\alpha/2)(\hat{\alpha} \times \vec{\tau}) \quad (14)$$

Simply, one can be determined  $|\vec{\Omega}|$  and  $|\vec{\Lambda}|$  as follows

$$\Omega = \tau\sqrt{\cosh^4(\alpha/2) - 2\cosh^2(\alpha/2)\sinh^2(\alpha/2)\cos 2\theta + \sinh^4(\alpha/2)} \quad (15)$$

$$\Lambda = 2\tau\sinh(\alpha/2)\cosh(\alpha/2)\sin\theta \quad (16)$$

where  $\theta$  is the angle between the local vectors  $\vec{\alpha}$  and  $\vec{\tau}$ . Again, for each section there are a local complex vector  $\vec{W}$  that may be determined depending on  $T_{PDL}$  and  $T_{PMD}$ . In other words,  $\vec{W}$  for each section does not depend on the other sections effect, but the total  $\vec{W}$  after  $N$  sections results from the contribution of all transmission matrices of the previous sections.

Accordingly, the following facts for the physical parameters of the individual sections

may be pointed. Generally, 1)  $\vec{\Omega} \cdot \vec{\Lambda} = 0$ , 2)  $\tau = \sqrt{\Omega^2 - \Lambda^2}$ , 3)  $\eta = 0$ , 4) the eigenvalues are real and 2)  $\cos\psi = \hat{p}_+ \cdot \hat{p}_- = \sinh^2\alpha \sin^2\theta - 1$ . That is; the

orthogonality is hold at each section if  $\alpha = 0$  or  $\theta = (0, \pi)$ . As such, the individual PDL effects will be discarded if no PDL or  $\vec{\alpha}$  and  $\vec{\tau}$  are parallel or anti- parallel. Also, if  $\alpha = 0$  or  $\theta = (0, \pi)$  then  $\Omega = \tau$  and  $\Lambda = 0$ , such that the DGD will not change, i.e.  $\tau_{old} = \tau_{new}$ . However, all fiber properties depend on the geometrical relation between the local vectors  $\vec{\alpha}$  and  $\vec{\tau}$  of each section.

### 3.2 Concatenation Sections Model

For  $N$  concatenation segments,  $T$  takes the form

$$T = T_N T_{N-1} T_{N-2} \dots T_2 T_1 = T_{PDLN} T_{PMDN} T_{PDLN-1} T_{PMDN-1} \dots T_{PDL2} T_{PMD2} T_{PDL1} T_{PMD1} \quad (17)$$

Now, the derivative and inverse of  $T$  are illustrated as

$$T_w = T_{PDLN} T'_{PMDN} T_{N-1} \dots T_2 T_1 + T_N T_{PDLN-1} T'_{PMDN-1} T_{N-2} \dots T_2 T_1 + \dots + T_N T_{N-1} \dots T_{PDL2} T'_{PMD2} T_1 + T_N T_{N-1} \dots T_2 T_{PDL1} T'_{PMD1} \quad (18)$$

$$T^{-1} = T_1^{-1} T_2^{-1} \dots T_{N-1}^{-1} T_N^{-1} \quad (19)$$

The frequency derivative of each  $T_{PMDj}$  may be computed using Eq.(3) to show

$$T'_{PMDj} = -\frac{i}{2} (\vec{\tau}_j \cdot \vec{\sigma}) T_{PMDj} \quad (20)$$

Substituting Eqs.(18) to (20) into (7), yields

$$\vec{W}_{tot} \cdot \vec{\sigma} = \vec{W}_N \cdot \vec{\sigma} + T_N (\vec{W}_{N-1} \cdot \vec{\sigma}) T_N^{-1} + T_N T_{N-1} (\vec{W}_{N-1} \cdot \vec{\sigma}) T_{N-1}^{-1} T_N^{-1} + \dots + T_N T_{N-1} \dots T_3 (\vec{W}_2 \cdot \vec{\sigma}) T_3^{-1} \dots T_{N-1}^{-1} T_N^{-1} + T_N T_{N-1} \dots T_3 T_2 (\vec{W}_2 \cdot \vec{\sigma}) T_2^{-1} T_3^{-1} \dots T_{N-1}^{-1} T_N^{-1} \quad (21)$$

where for each isolated section, we have  $\vec{W}_j \cdot \vec{\sigma} = T_{PDLj} (\vec{\tau}_j \cdot \vec{\sigma}) T_{PDLj}^{-1}$ . Moreover, Eq.(21) may be simplified further using the assumption  $\vec{T}_{j+1} = T_o T_N T_{N-1} \dots T_{j+1}$  to get

$$\vec{W}_{tot} \cdot \vec{\sigma} = \sum_{j=1}^N \vec{T}_{j+1} (\vec{W}_j \cdot \vec{\sigma}) \vec{T}_{j+1}^{-1} \quad (22)$$

Each term in the last equation represents  $T_w T^{-1}$  for one or more sections. Therefore, each term represents a traceless matrix, which can be written as  $\vec{a}_j \cdot \vec{\sigma}$ . Here  $\vec{a}_j$  computed as  $\vec{a}_j = \text{trace}(\vec{\sigma} \vec{T}_{j+1} (\vec{W}_j \cdot \vec{\sigma}) \vec{T}_{j+1}^{-1}) / 2$ . As such, the total PMD vector may be expressed as

$$\vec{W}_{tot} = \sum_{j=1}^N \vec{a}_j = \frac{1}{2} \sum_{j=1}^N \text{trace}(\vec{\sigma} \vec{T}_{j+1} (\vec{W}_j \cdot \vec{\sigma}) \vec{T}_{j+1}^{-1}) \quad (23)$$

Also, Eq.(21) may be reformed to obtain the following recursive relation

$$\vec{W}^{(N)} \cdot \vec{\sigma} = \vec{W}_N \cdot \vec{\sigma} + T_N (\vec{W}_{N-1} \cdot \vec{\sigma}) T_N^{-1} \quad (24)$$

where  $\vec{W}^{(N)}$  is the PMD vector of the total N segments, whereas  $\vec{W}_N$  is the PMD vector of the N-th segment only. As an example to determine  $\vec{W}^{(N)}$ , it is clear that:

$$\vec{W}^{(0)} = 0, \quad \vec{W}^{(1)} = \vec{W}_1, \quad \vec{W}^{(2)} = \vec{W}_2 + \frac{1}{2} [\vec{\sigma} T_2 (\vec{W}_1 \cdot \vec{\sigma}) T_2^{-1} + T_2 (\vec{W}_1 \cdot \vec{\sigma}) T_2^{-1} \vec{\sigma}]$$

and so on, where the identities in Eqs.(12) will be used to extract  $\vec{W}$  from  $\vec{W} \cdot \vec{\sigma}$ . It is important to note that, the diagonal elements in the terms

$[\vec{\sigma} T_N (\vec{W}_{N-1} \cdot \vec{\sigma}) T_N^{-1} + T_N (\vec{W}_{N-1} \cdot \vec{\sigma}) T_N^{-1} \vec{\sigma}]$  are identical. Eqs.(23) and (24) represent new recursive formulas of the complex PMD vector, which give the idea about the amount of difficulties to obtain a closed form of output PMD vector of the pulse propagates through optical fiber.

#### 4. Results and Discussions

We are used Eq.(23) to calculate the parameters that are effected during evolution of light through an optical fiber. At simulation, we are used 100000 fibers, each with long 100 km and varying  $D_p$  (PMD) and  $D_\alpha$  (PDL). Each fiber will be divided to 100 sections where the random matrices for each section is generated using the relations at Eqs.(2) and (3). The present results represent the average over the 100000 fibers. It is important to note that, the average is the nearest to the practical behavior more than a single simulation, where the larger number of fibers will make the results to be very accurate.

Fig.(1) represent the orthogonality of PSP of a system with varying  $D_p$  and  $D_\alpha$ . According to Eq. (11) the PSP of a system are not necessarily to be orthogonal any more, in other words, the angle between two PSPs in stokes space is not necessarily  $180^\circ$ , our simulation results in Fig.(1) show that. In case of very low PDL value (blue curve in each subfigure) the angle is almost  $90^\circ$  (in jones space), with an increasing rate of PDL, more and more the PSP well not to be orthogonal any more, the angles may reach down below  $45^\circ$  in same regions, so in this Fig. we can see the variation of PMD will be induced a very small change in the probability of orthogonality.

Fig.(2) represent the probability density function (pdf) for the real DGD of the system with varying PMD and PDL, we can see in this Fig. the real DGD ( $\tau$ ) is Maxwellian distributed just in systems without any PDL, in system with very high PDL the distribution of the real part DGD may change reasonably with increased possibility of high DGD values and the distribution closely to be Gaussian Maxwellian with increasing PMD and PDL, so in this Fig. we can see in subfigure when high PDL the pdf of zero DGD will be increasing. The statistics of the real part of DGD depend on the values of  $D_p$  (PMD) and  $D_\alpha$  (PDL) but PMD effected on real DGD more than PDL.

Fig.(3) represent the probability density function (pdf) for the imaginary part of DGD or we can called DSA of the system with varying PMD and PDL, this Fig. show the DSA is ( $\text{sech}^2$ ) distributed when PMD and PDL are small values, but when increasing the values of PMD and PDL its clearly that the pdf of DSA well be change, so DSA depend on both PMD and PDL.

Finally, the PSPs are orthogonal at  $PDL=0$ , while the presence of any PDL value will eliminate the orthogonality. The amount of orthogonality deviation depends on the

PDL value, where the larger PDL causes the smaller  $\psi$ . In other words, decreasing PDL will make the PSPs near the anti-parallel. Generally, the presence of PMD causes DGD and peak shifting but the PSP's orthogonality will maintain. On the other hand, the impact of PDL will increase the mean value of DGD (more than the case of no PDL) and does not reserve the PSPs orthogonality. In turn, this will cause distortion in the resulted pulse shape and consequently the degree of polarization will be reduced. All changes are subject to increase the number of sections (length) and PDL value. Fluctuations in all curves become smoother by increasing the number of optical fibers used in the simulation.

## 5. Conclusions

In conclusion, the proposed treatments to study the randomness behavior in presence of PMD and PDL gave good results. Where it is possible to simulate the optical fiber through the knowledge of PDL,  $D_p$  and the generated local vectors to infer the fluctuations of  $\tau$ ,  $\tau_{eff}$ ,  $\eta$ ,  $\vec{\Omega} \cdot \vec{\Lambda}$  and  $\cos\psi$  through evolution. In general, we can work close to the case of orthogonal PSPs if PDL is very small, whereas the change of PMD does not affect the orthogonality.

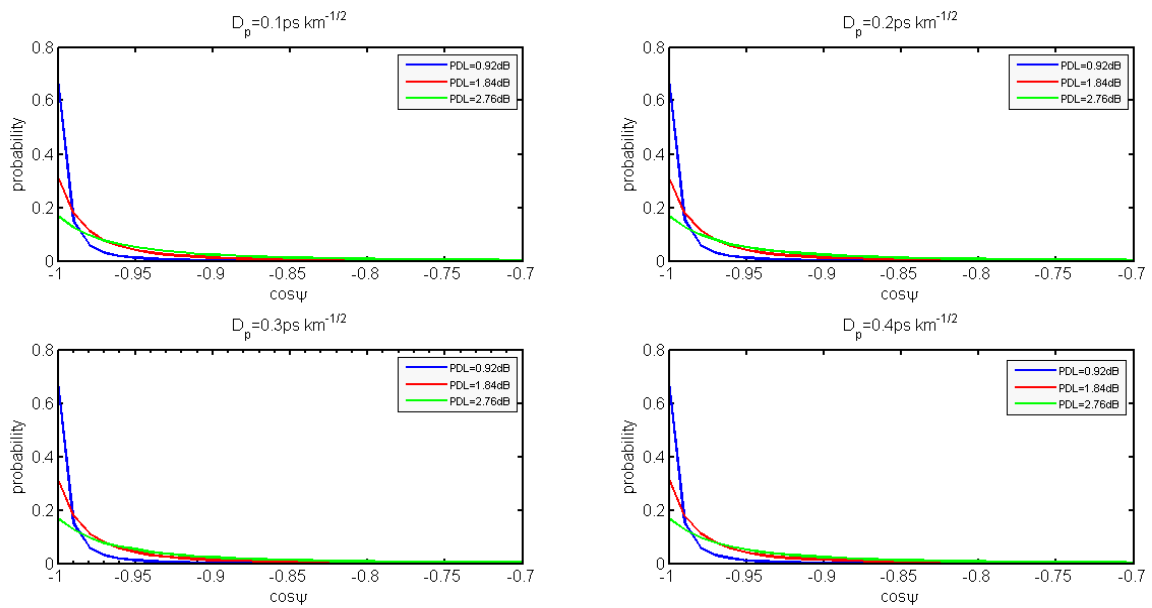


Fig.(1): Orthogonality of PSP of a system with varying PMD and PDL, where the three curves (blue, red and green) represent the orthogonal probability of PSP for PDL=(0.92, 1.84 and 2.76)dB respectively

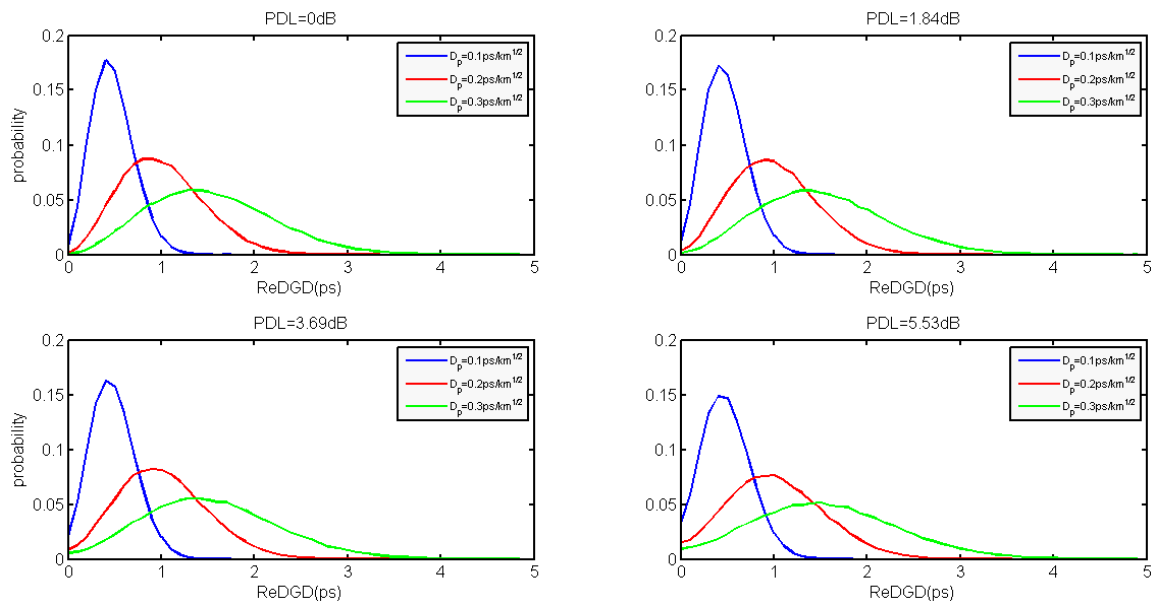


Fig.(2): pdf for real DGD of a system with varying PDL and PMD, where the three curves (blue, red and green) represent the pdf for DGD for  $D_p=(0.1, 0.2$  and  $0.3)$ ps/sqrt(km) respectively

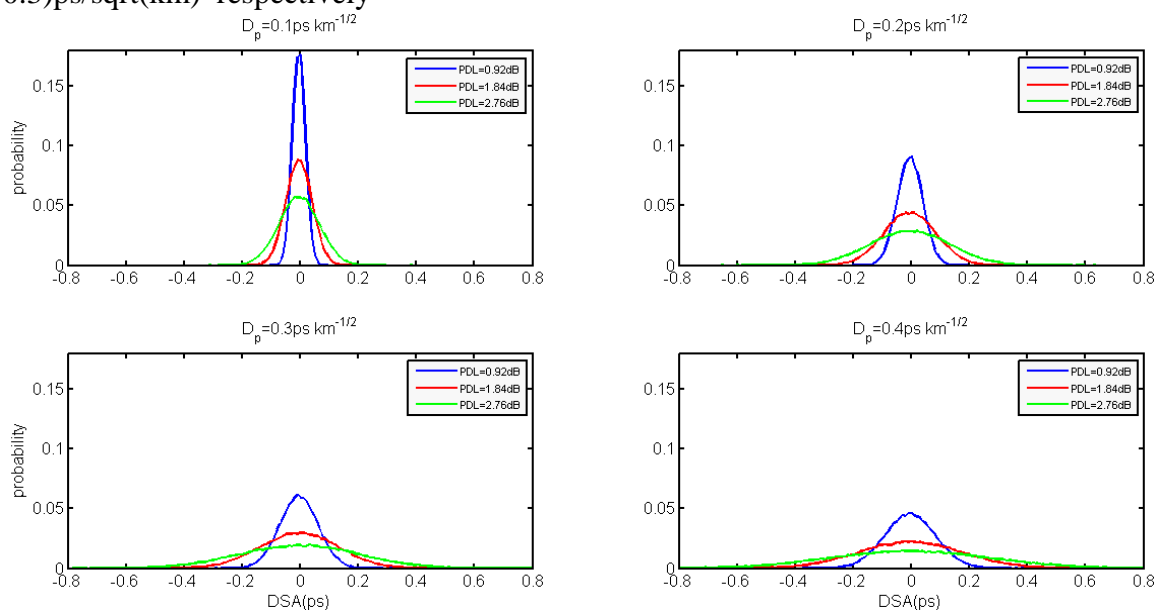


Fig.(3): pdf for DSA of a system with varying PMD and PDL, where the three curves (blue, red and green) represent the pdf for DSA for  $PDL=(0.92, 1.84$  and  $2.76)$ dB respectively

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